
ELE5481

**SIGNAL PROCESSING OPTIMIZATION
TECHNIQUES**

5. LINEAR AND QUADRATIC PROGRAMS

Linear Programming (LP)

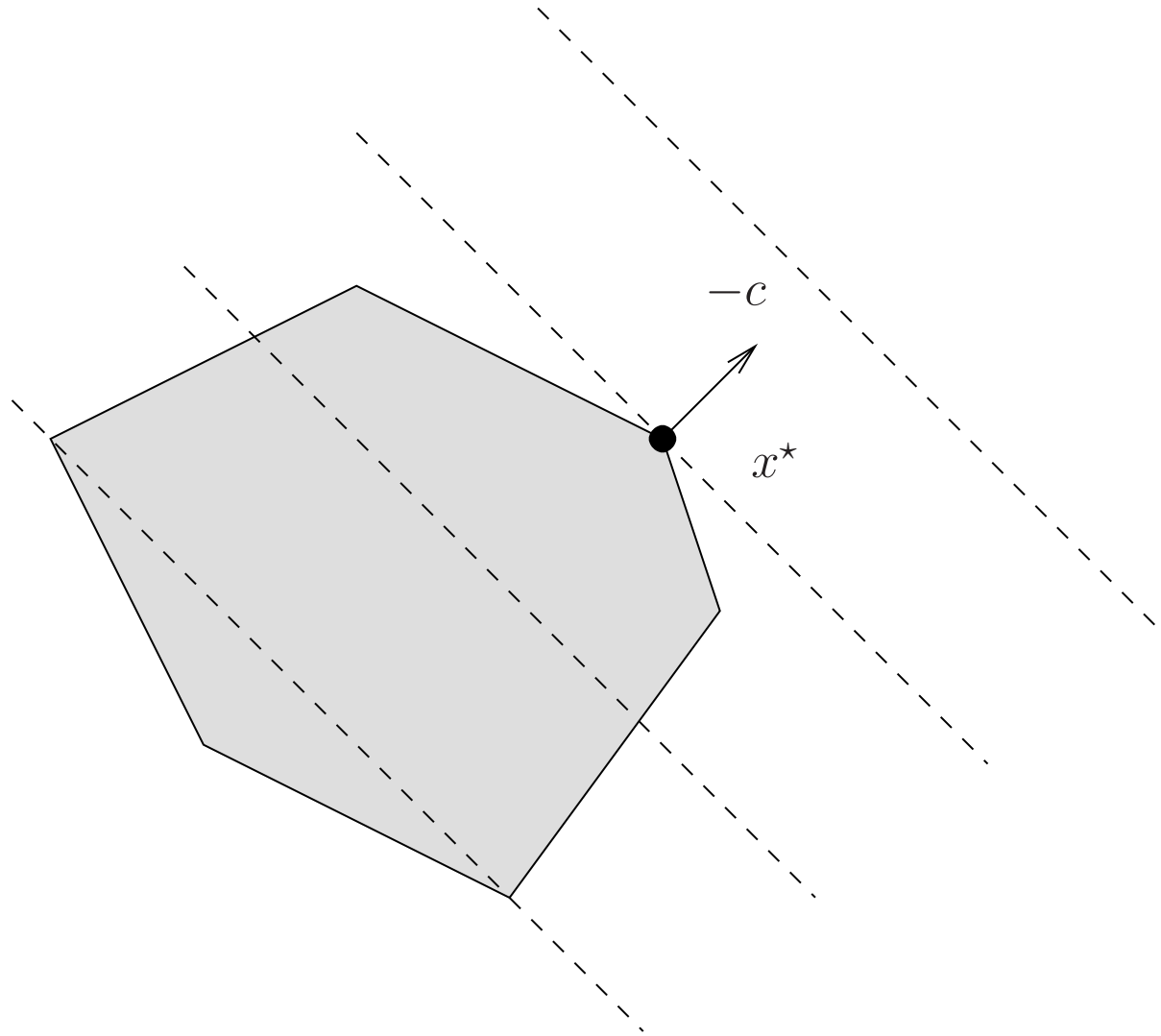
A general form of LP:

$$\begin{aligned} \min \quad & c^T x \\ \text{s.t.} \quad & Gx \preceq h \\ & Ax = b \end{aligned} \tag{1}$$

A standard form of LP widely used in the literature & software:

$$\begin{aligned} \min \quad & c^T x \\ \text{s.t.} \quad & x \succeq 0 \\ & Ax = b \end{aligned} \tag{2}$$

LP is a problem of minimizing a linear objective function over a polyhedron.



The general form in (1) can be reformulated as the standard form in (2). Problem (1) is equiv. to

$$\begin{aligned} \min \quad & c^T x \\ \text{s.t.} \quad & s \succeq 0, \quad h - Gx = s, \quad Ax = b \end{aligned} \quad (*)$$

Let $x = x^+ - x^-$, where $x^+, x^- \succeq 0$. Eq. (*) is equiv. to

$$\begin{aligned} \min \quad & [c^T \quad -c^T \quad 0] \begin{bmatrix} x^+ \\ x^- \\ s \end{bmatrix} \\ \text{s.t.} \quad & \begin{bmatrix} x^+ \\ x^- \\ s \end{bmatrix} \succeq 0, \quad \begin{bmatrix} A & -A & 0 \\ G & -G & I \end{bmatrix} \begin{bmatrix} x^+ \\ x^- \\ s \end{bmatrix} = \begin{bmatrix} b \\ h \end{bmatrix} \end{aligned}$$

A Brief History

- 1939: planning, production (Kantorovich)
 - Noble prize in Economics, 1975
- 1940's: simplex algorithm for LP (Dantzig)
 - used in Berlin airlift, 1948
- 1970's: polynomial-time ellipsoid algorithm for LP (Khachiyan)
 - based on work by Dirkin, Shor and Nemirovski in 1960's
 - front page news in the Western world, incl. New York Times (exaggerated)
- 1980's: polynomial-time interior-point algorithm for LP (Karmarkar)
- late 1980's-now: polynomial-time interior-point methods for nonlinear convex programs (Nesterov and Nemirovski, 1994)
 - convex opt. software we used today is largely based on interior-point methods.

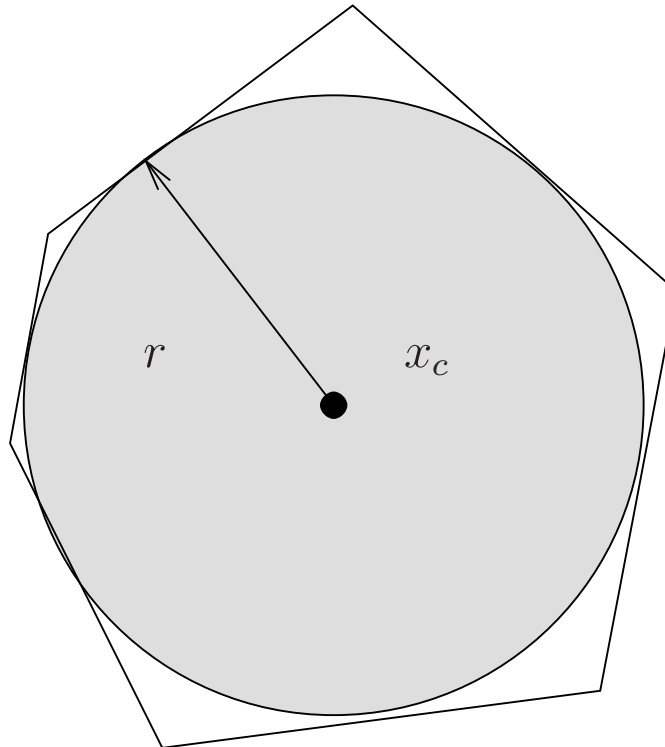
Application: Diet Problem

- x_i is the quantity of food i .
- Each unit of food i has a cost of c_i .
- One unit of food j contains an amount A_{ij} of nutrient j .
- We want nutrient i to be at least equal to b_i .
- **Problem:** find the cheapest diet such that the minimum nutrient requirements are fulfilled.
- This problem can be cast as an LP:

$$\begin{aligned} \min \quad & c^T x \\ \text{s.t.} \quad & Ax \preceq b, \quad x \succeq 0 \end{aligned}$$

Chebychev Center

- Let a norm ball $B(x_c, r) = \{ x \mid \|x_c - x\|_2 \leq r \}$, & a polyhedron $\mathcal{P} = \{ x \mid a_i^T x \leq b_i, i = 1, \dots, m \}$.
- **Problem:** Find the largest ball inside a polyhedron \mathcal{P} ; i.e., $\max_{x_c, r} r$, subject to $B(x_c, r) \subseteq \mathcal{P}$.



- An alternative representation of the norm ball: $B(x_c, r) = \{ x_c + u \mid \|u\|_2 \leq r \}$.

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$$B(x_c, r) \subseteq \mathcal{P} \iff \sup_u \{ a_i^T (x_c + u) \mid \|u\|_2 \leq r \} \leq b_i, \quad \forall i$$

$$\iff a_i^T x_c + r \|a_i\|_2 \leq b_i, \quad \forall i$$

- Hence, the Chebychev center problem is equiv. to an LP

$$\max r$$

$$\text{s.t. } a_i^T x_c + r \|a_i\|_2 \leq b_i, \quad i = 1, \dots, m$$

Piecewise Linear Minimization

$$\min \max_{i=1,\dots,m} (a_i^T x + b)$$

By using the epigraph form, the problem is equiv. to

$$\min t$$

$$\text{s.t. } \max_{i=1,\dots,m} (a_i^T x + b) \leq t$$

$$\iff \min t$$

$$\text{s.t. } a_i^T x + b \leq t, \quad i = 1, \dots, m$$

which is an LP.

ℓ_∞ -norm (Chebychev) Approximation

$$\min \|Ax - b\|_\infty$$

Using the epigraph form, the ℓ_∞ -norm approx. problem can be cast as an LP:

$$\min t$$

$$\text{s.t. } \max_{i=1,\dots,m} |r_i| \leq t$$

$$r = Ax - b$$

$$\iff \min t$$

$$\text{s.t. } -t\mathbf{1} \preceq r \preceq t\mathbf{1}$$

$$r = Ax - b$$

ℓ_1 -norm Approximation

$$\min \|Ax - b\|_1$$

can be rewritten as

$$\begin{aligned} & \min \sum_{i=1}^m |r_i| \\ & \text{s.t. } r = Ax - b \\ \iff & \min \sum_{i=1}^m t_i \\ & \text{s.t. } -t_i \leq r_i \leq t_i, \quad i = 1, \dots, m \\ & r = Ax - b \end{aligned}$$

which is an LP.

Linear Fractional Programming

$$\begin{aligned} \min \quad & \frac{c^T x + d}{f^T x + g} \\ \text{s.t.} \quad & Ax \preceq b \\ & f^T x + g > 0 \end{aligned}$$

- The objective function is quasiconvex, and each of its sublevel sets is a polyhedron.

If the feasible set $\{x | Ax \preceq b, f^T x + g > 0\}$ is bounded, the linear fractional program can be transformed to an equiv. LP:

$$\begin{aligned} \min_{y \in \mathbf{R}^n, z \in \mathbf{R}} \quad & c^T y + dz \\ \text{s.t.} \quad & Ay - bz \preceq 0, \quad z \geq 0 \\ & f^T y + gz = 1 \end{aligned}$$

- It can be shown that if $\{x | Ax \preceq b, f^T x + g > 0\}$ is bounded, then $z > 0$ for any feasible (y, z) .
- If (y, z) is feasible in the LP, then $x = y/z$ is feasible in the linear fractional program.
- This transformation is known as the Charnes-Cooper transformation.

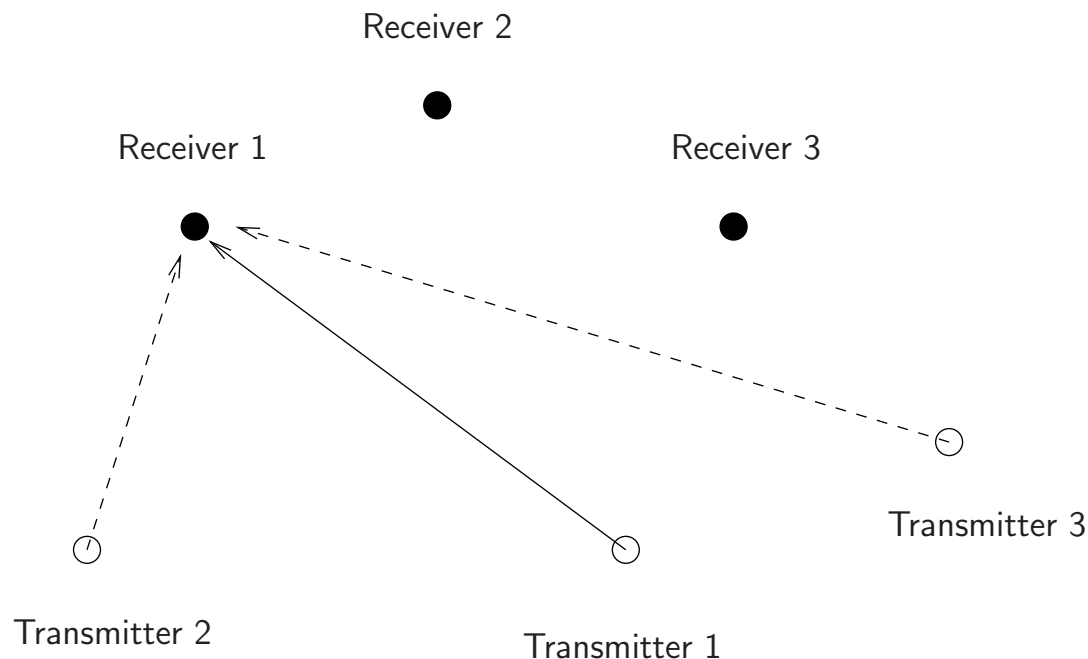
Generalized linear fractional program:

$$\begin{aligned} \min \quad & \max_{i=1, \dots, K} \frac{c_i^T x + d}{f_i^T x + g_i} \\ \text{s.t.} \quad & Ax \preceq b \\ & f_i^T x + g_i > 0, \quad i = 1, \dots, K \end{aligned}$$

- The objective function is quasiconvex.
- Can be solved using the bisection method.

Example: Optimal Power Assignment

- K transmitters, K receivers.



- Transmitter i sends signals to receiver i , & the other transmitters are interferers.
- The signal-to-interference-and-noise ratio (SINR) at receiver i

$$\gamma_i = \frac{G_{ii}p_i}{\sum_{j \neq i} G_{ij}p_j + \sigma_i^2}$$

where p_i is the transmitter i power, G_{ij} is the path gain from transmitter j to receiver i , and σ_i^2 is the noise power at receiver i .

- **Problem:** Maximize the smallest γ_i subject to power constraints $0 \leq p_i \leq p_{\max,i}$, where $p_{\max,i}$ is the max. allowable power of transmitter i .

The power assignment problem is

$$\max_{\substack{p_i \in [0, p_{\max,i}] \\ i=1, \dots, K}} \min_{i=1, \dots, K} \frac{G_{ii}p_i}{\sum_{j \neq i} G_{ij}p_j + \sigma_i^2}$$

which can be reformulated as a generalized linear fractional program:

$$\begin{aligned} \min \quad & \max_{i=1, \dots, K} \frac{\sum_{j \neq i} G_{ij}p_j + \sigma_i^2}{G_{ii}p_i} \\ \text{s.t.} \quad & 0 \leq p_i \leq p_{\max,i}, \quad i = 1, \dots, K \end{aligned}$$

- Note: The power assignment problem can alternatively be solved by geometric programming, or by a closed-form method that utilizes the problem structure $G_{ij} \geq 0, \sigma_i^2 > 0$.

Example: Another Optimal Power Assignment Problem

Problem: Minimize the average transmitter power, subject to a constraint that all SINRs are not less than a pre-specified threshold γ_o .

$$\begin{aligned} \min_p \quad & \sum_{i=1}^K p_i \\ \text{s.t.} \quad & \frac{G_{ii}p_i}{\sum_{j \neq i} G_{ij}p_j + \sigma_i^2} \geq \gamma_o, \quad i = 1, \dots, K \\ & p_i \geq 0, \quad i = 1, \dots, K \end{aligned}$$

- The problem can be rewritten as

$$\begin{aligned} \min_p \quad & \sum_{i=1}^K p_i \\ \text{s.t.} \quad & -G_{ii}p_i + \gamma_o \sum_{j \neq i} G_{ij}p_j + \gamma_o \sigma_i^2 \leq 0, \quad i = 1, \dots, K \\ & p_i \geq 0, \quad i = 1, \dots, K \end{aligned}$$

which is an LP.

- Note: There is a closed-form solution to the LP above, when taking into account the problem structure $G_{ij} \geq 0$, $\sigma_i^2 > 0$.

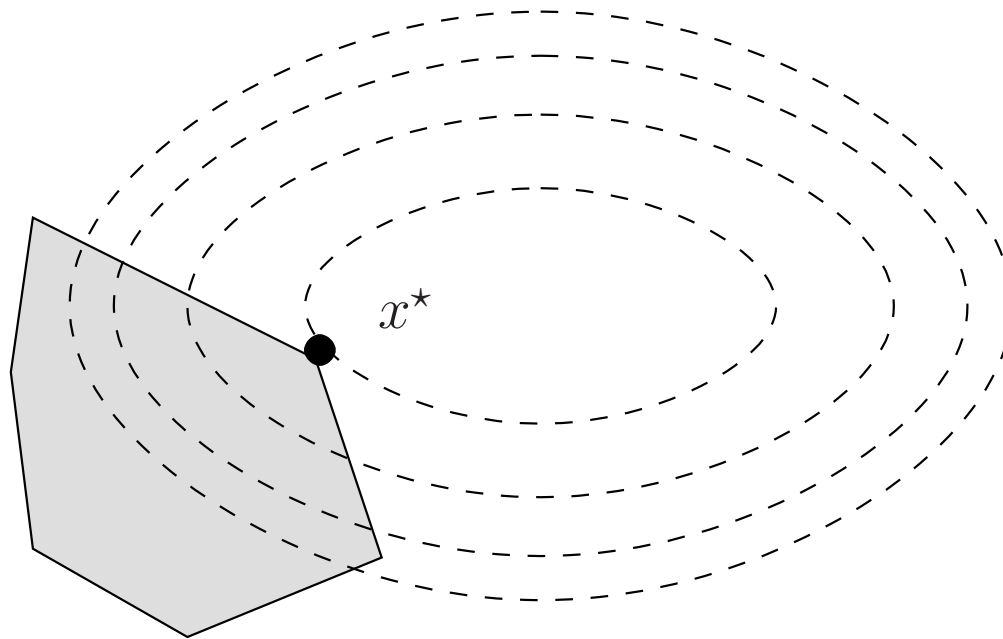
Additional Reading

- [SB04]** M. Schubert and H. Boche, “Solution of the multiuser downlink beamforming problem with individual SINR constraints,” *IEEE Trans. Vehicular Tech.*, 2004.

Quadratic Programming (QP)

$$\begin{aligned} \min \quad & \frac{1}{2}x^T P x + q^T x + r \\ \text{s.t.} \quad & Ax = b, \quad Gx \preceq h \end{aligned}$$

A QP is convex iff $P \succeq 0$.



- Unconstrained QP (or LS)

$$\min \frac{1}{2}x^T P x + q^T x + r$$

is a special case of QP where a closed form solution is available.

- The optimality condition is $Px = -q$.
 - If $P \succ 0$ then $x^* = -P^{-1}q$.
 - If $P \succeq 0$ but $q \notin \mathcal{R}(P)$, then there is no solution for $Px = -q$. It can also be shown that $p^* = -\infty$.
 - If $P \succeq 0$ & $q \in \mathcal{R}(P)$, then $x^* = -P^\dagger q + \nu$ for any $\nu \in \mathcal{N}(P)$.

Examples:

- LS with bound constraints:

$$\begin{aligned} \min \quad & \|Ax - b\|_2^2 \\ \text{s.t.} \quad & \ell \preceq x \preceq u \end{aligned}$$

- Distance between polyhedra:

$$\begin{aligned} \min \quad & \|x_1 - x_2\|_2^2 \\ \text{s.t.} \quad & x_1 \in \{x \mid A_1 x \preceq b_1\}, \quad x_2 \in \{x \mid A_2 x \preceq b_2\} \end{aligned}$$

$$\iff \begin{aligned} \min \quad & \|x_1 - x_2\|_2^2 \\ \text{s.t.} \quad & A_1 x_1 \preceq b_1, \quad A_2 x_2 \preceq b_2 \end{aligned}$$

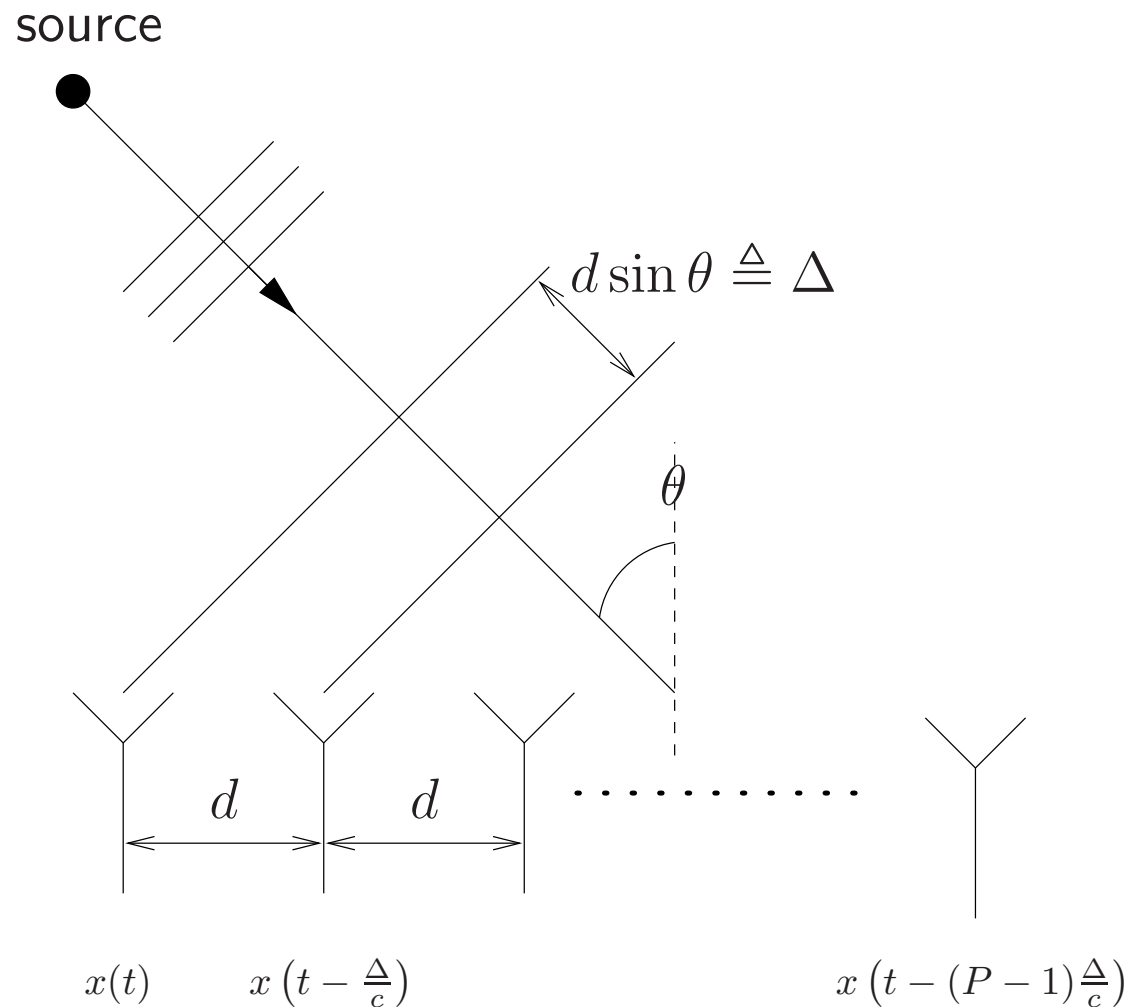
Quadratically Constrained QP (QCQP)

$$\begin{aligned} \min \quad & \frac{1}{2}x^T P_0 x + q_0^T x + r_0 \\ \text{s.t.} \quad & \frac{1}{2}x^T P_i x + q_i^T x + r_i \leq 0, \quad i = 1, \dots, m \\ & Ax = b \end{aligned}$$

- QCQP is convex if $P_i \succeq 0$ for all i .
- When $P_i \succ 0$ for $i = 1, \dots, m$, QCQP is a quadratic min. problem over an intersection of ellipsoids.
- If $P_i = 0$ for $i = 1, \dots, m$, then QCQP reduces to QP.
- If $P_i = 0$ for $i = 0, 1, \dots, m$, then QCQP reduces to LP.

Beamformer Design via QPs

Uniform linear array:



Signal model:

A1) far-field situations so that source waves are planar; &

A2) narrowband source signals so that the received signal of one sensor is a phase shifted version of that of another.

If a source signal $s(t) \in \mathbf{C}$ comes from a direction of θ , the array output $y(t) = [y_1(t), \dots, y_P(t)]^T$ is

$$y(t) = a(\theta)s(t)$$

Here,

$$a(\theta) = [1, e^{-j2\pi d \sin(\theta)/\lambda}, \dots, e^{-j2\pi d(P-1) \sin(\theta)/\lambda}]^T \in \mathbf{C}^P$$

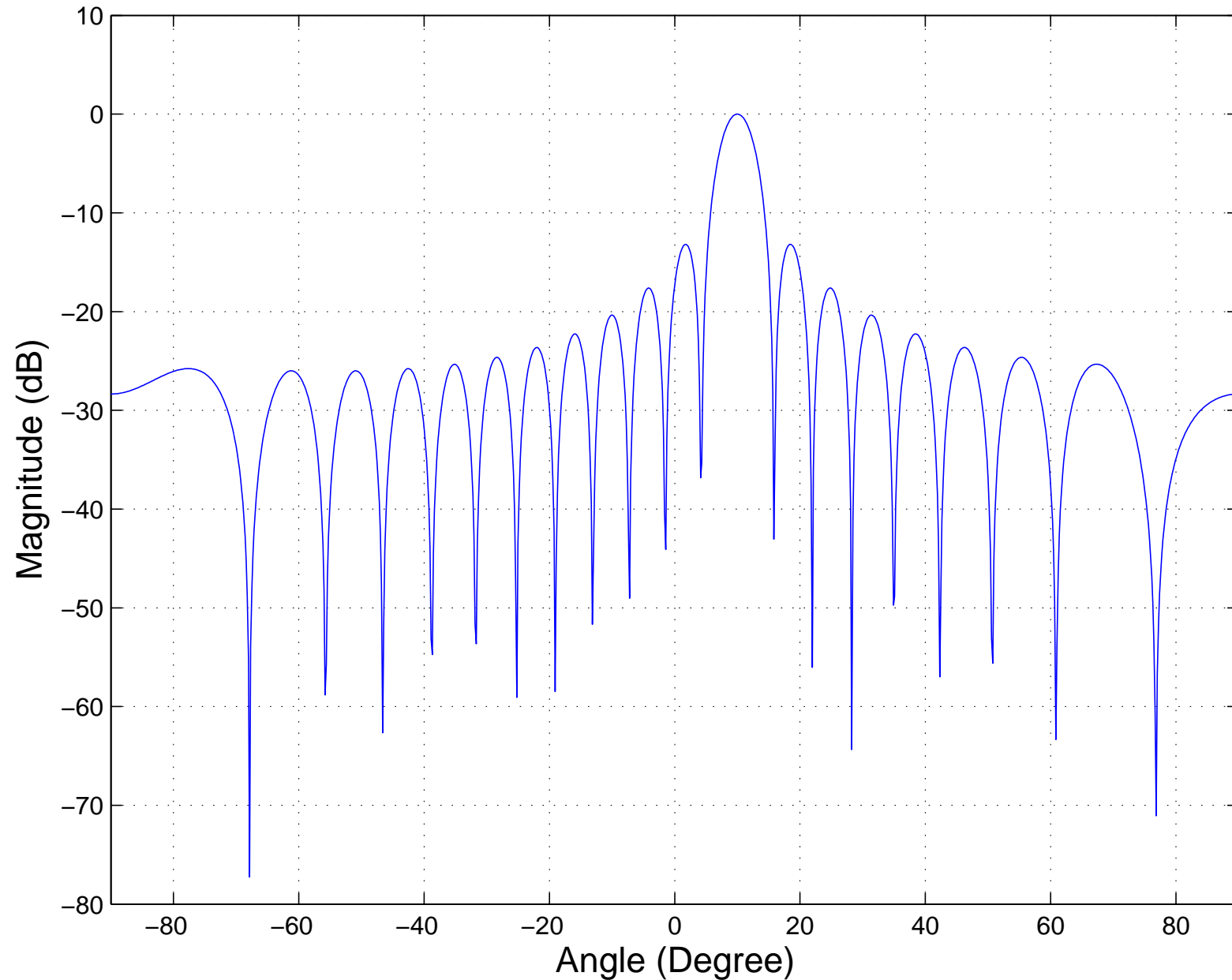
is the **steering vector**, where λ is the signal wavelength.

Beamforming:

$$\hat{s}(t) = w^H y(t)$$

where $w \in \mathbf{C}^P$ is a beamformer weight vector.

- Let $\theta_{\text{des}} \in [-\frac{\pi}{2}, \frac{\pi}{2}]$ be the desired direction.
- A simple beamformer is $w = a(\theta_{\text{des}})$, but it does not provide good sidelobe suppression.
- **Problem:** find a w that minimizes sidelobe energy subject to a pass response to θ_{des} .



Direction pattern of the conventional beamformer. $\theta_{\text{des}} = 10^\circ$; $P = 20$.

- Let $\Omega = [-\frac{\pi}{2}, \theta_\ell] \cup [\theta_u, \frac{\pi}{2}]$ denote the sidelobe band, for some θ_ℓ, θ_u so that $\theta_{\text{des}} \in [\theta_\ell, \theta_u]$.
- **Average sidelobe energy minimization:**

$$\begin{aligned} \min \quad & \int_{\Omega} |w^H a(\theta)|^2 d\theta \\ \text{s.t.} \quad & w^H a(\theta_{\text{des}}) = 1 \end{aligned}$$

The problem is equiv. to an equality constrained QP:

$$\begin{aligned} \min \quad & w^H P w \\ \text{s.t.} \quad & w^H a(\theta_{\text{des}}) = 1 \end{aligned}$$

where $P = \int_{\Omega} a(\theta) a^H(\theta) d\theta$ (can be computed by numerical integration).

- **Worst-case sidelobe energy minimization:**

$$\begin{aligned} \min \quad & \max_{\theta \in \Omega} |w^H a(\theta)|^2 \\ \text{s.t.} \quad & w^H a(\theta_{\text{des}}) = 1 \end{aligned}$$

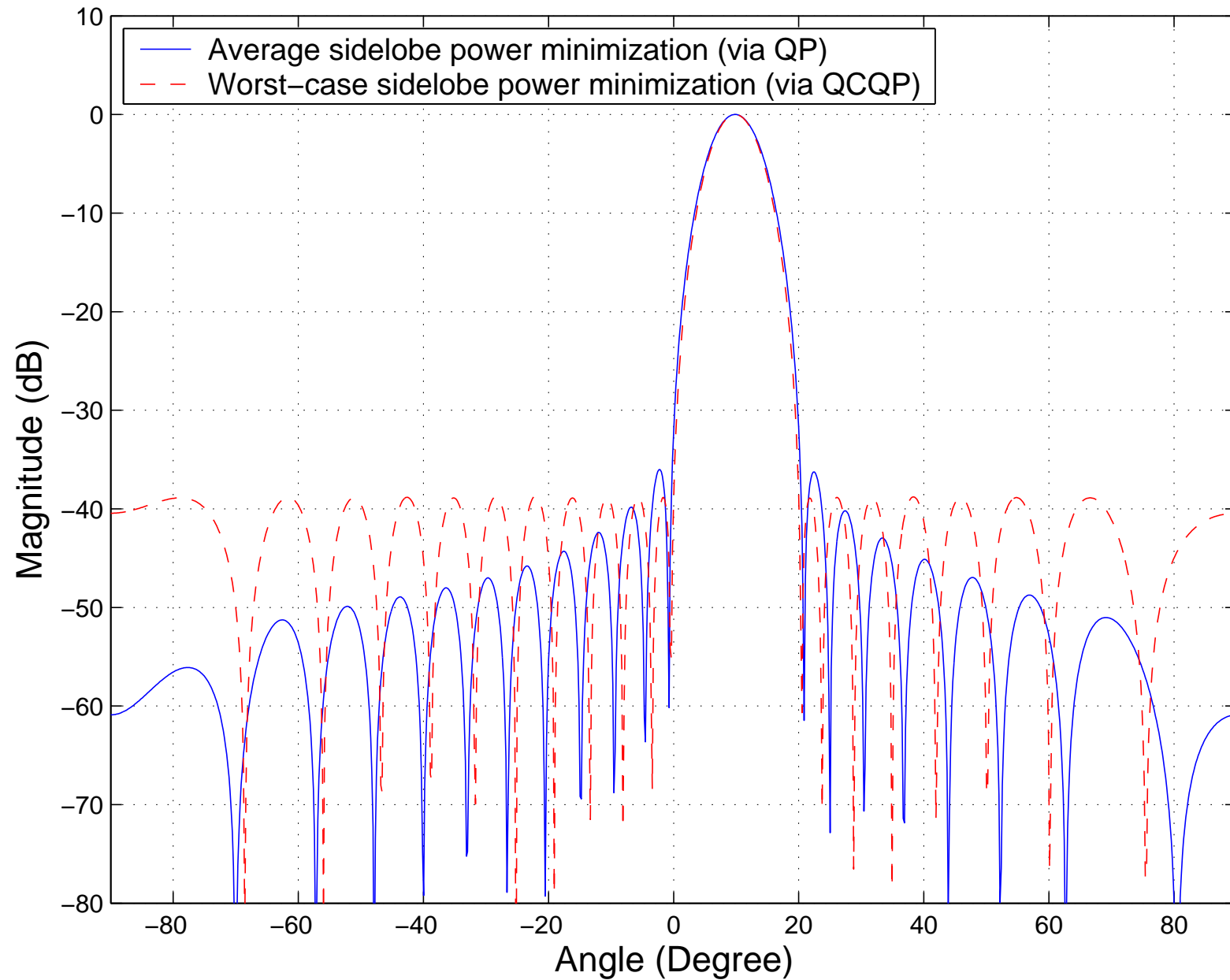
The problem can be reformulated as

$$\begin{aligned} \min \quad & t \\ \text{s.t.} \quad & |w^H a(\theta)|^2 \leq t, \quad \theta \in \Omega \\ & w^H a(\theta_{\text{des}}) = 1 \end{aligned}$$

which is a QCQP with semi-infinite constraints.

- The worst-case sidelobe energy minimization problem can be approximated by discretization.
- Let $\theta_1, \theta_2, \dots, \theta_L$ be some set of sample points in Ω . We approximate the problem by

$$\begin{aligned} \min t \\ \text{s.t. } |w^H a(\theta_i)|^2 \leq t, \quad i = 1, \dots, L \\ w^H a(\theta_{\text{des}}) = 1 \end{aligned}$$



Direction patterns of the two beamformer designs. $\theta_{\text{des}} = 10^\circ$. Sidelobe suppression is applied to directions outside $[0^\circ, 20^\circ]$.